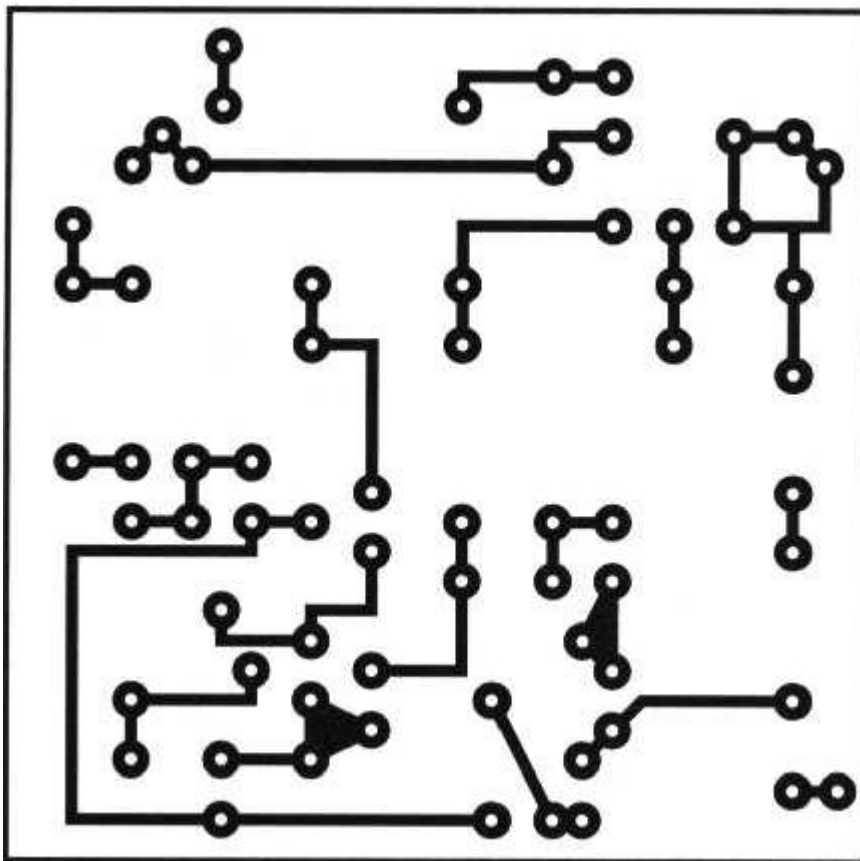


The parts used are common and easy to get, no miniaturising and a lot of space on the PCB-layout, so that the use of parts from the junkbox with other dimensions could be guaranteed. The shielded-boxes were bought, but they also could be made of PCB-material. The first version of this transverter was build in deadbug- design, but to make things easier we developed PCB's.

The PCB's are doublesided and the Rx/Tx units use the component side only for shielding and ground connections of components. It is important to clear the drilled holes from the component side so that no shortcuts are manufactured when soldering the pass through component-connections on the etched side. The components that have to be connected to (ground / shield) are marked on the layout picture. This type of design has a lot of advantages, you have a good RF-shielding to the layoutside and it is easy possible to shield units from each other by putting walls between them.

The two amplifiers and the filter were build in an other way. We also used doublesided PCB's but the components were placed on the layoutside, because it is easier to handle the cooling of the RF-transistors in this way.

RX/TX switching and power supply can be buildup as one prefers.



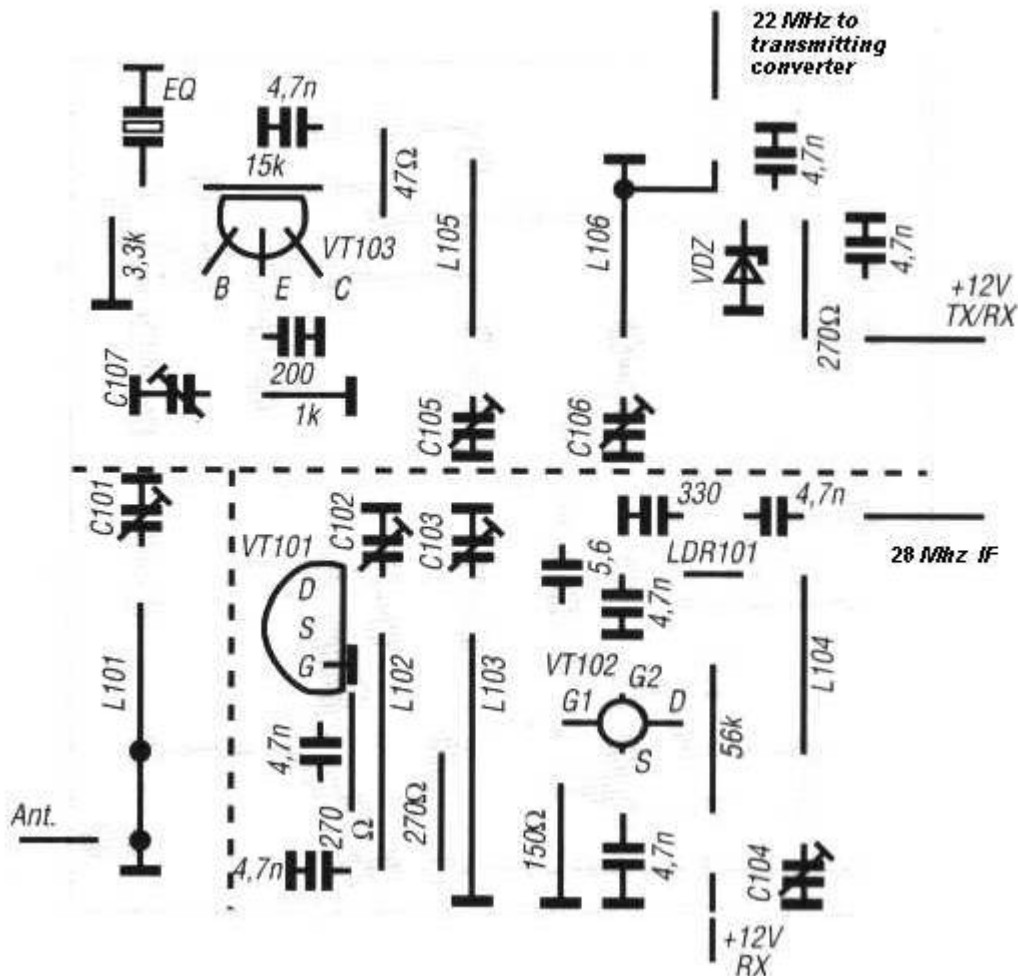
The lokal oscillator (Lo)

Quarz oscillator and reciving-converter are together on one PCB. The power supply connections to the units are seperated because the reciving-converter must be switched off in transmit mode as the Lo keeps up operation in order to supply the TX-mixer with a 22 MHz (24MHz) signal 0.3 Vp-p

The oscillator is build with a cheap 22 or 24 MHz quartz in a HC-18U case. The highgrade circuit is necessary so that it can be guaranteed that the quartz swings on. Even low quality quartz from coputers can be pulled to 22 or 24MHz by tuning L105 and C107.

The output signal of the oscillator is then passed through a bandpass filter (C106 and L106) to reduce harmonics and spurious products. If it wouldn't have been done in this way you would have problems with doubling the frequency to 44MHz with a much too high power to the input of the mixer, producing a very spurious signal.

The signal is being capacitively picked up over a 5.6pF Cap. at high impedance from the second resonant stage of the bandpass-filter and fed to the mixer in the receiving-converter. The signal for the transmitting-converter is picked up at a tap on the cold side of L106. This stage is always supplied with 12 V DC and has a zener of 9.1V to stabilise the voltage.



The receiving-converter

We used a low noise FET typ BF256 C (VT101) in grounded gate at the input to cancel self-oscillation. The input-stage will work stable even if the antenna is not attached. After the FET, the signal passes through a bandpass and is mixed with help of a dual-gate-mosfet-stage BF961 (VT102). The RX-signal is fed to gate1, and the LO to gate 2. This type of mixing is quite common and is being used since the last 10 years for these cases.

The drain is connected to a tuned Pi-filter (L104) to a frequency between 26 and 30MHz with a bandwidth of about 1 MHz depending on the bandwidth and Frequency wanted. The obtained noise factor of about 2 dB is actually the best you can get. It isn't necessary investing more time in better performance because the noise figure is always quite high on this band.

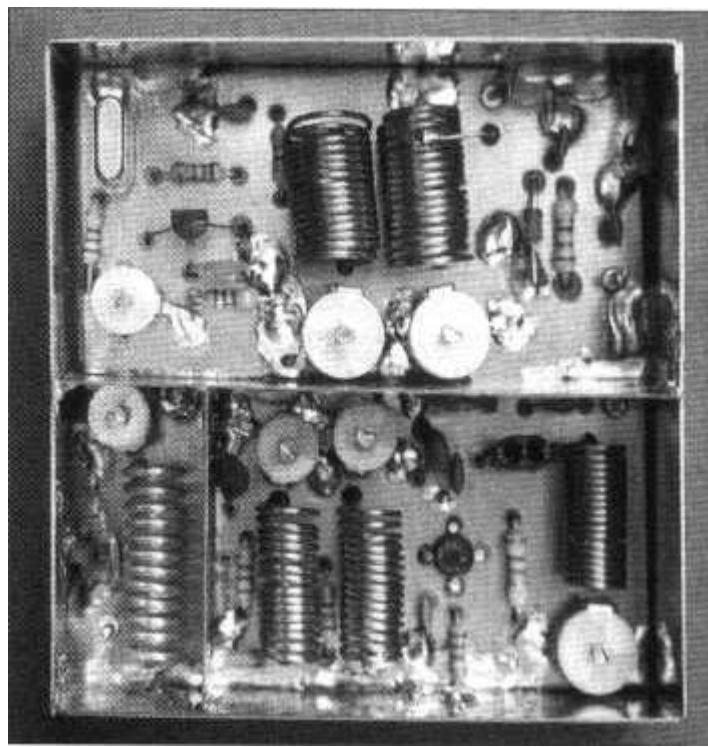
Building the oscillator and reciving-converter.

At first you should get accostomed to the connections of the transistors (see figure 2), in order not to make mistakes. Logically we start with the reciving-converter and the lokal-oscillator. The dimensions of the PCB is 71mm X 71mm (abt. 2.79 inches in square), and would fit in a standard tinned metal case of 72mm X 72mm.

At first you set the coils , all the coils are airwound ones, look at figure 4. The coils are closewound on drillshafts with a certain amount of windings and after that pulled appart to fit the drilled holes on the PCB (L101, L102 and L103). all the other coils are closewound. On the component layout-plan you also see the marks to put up the RF-shielding walls. The cover of the unit must have holes at particular areas to be capebel of tuning the unit when the cover is set on it. The other details can bee seen by studying the picture. (figure 5)

Adjusting the LO / reciving-converter unit

To make the adjusting of the unit a bit easier you can preconfigure the capasitors to the later mentioned positions, you should do this in anyway, Tuning the capacitance compleatly out or in is a sign for an error! At first give power to the oscillator only, with help of a griddip or reciver tuned to the oscillator frequency (22 or 24MHz), tune to maximum reading on the s-meter, after that tune the oscillator to the exact frequency by adjusting the paralell capacitance to the quarz. Here you should pay attention to the frequency and also to the safe initial start of oscillation as you put power to it. Now you can connect the reciving-converter to the exciter and fix the output circuit of the 28MHz (26MHz) IF to maximum noisfigure in the exciter by tuning C104, then tune C103 and C102. Attach a 6m antenna to the circuit, you must hear a change to the noise as yo do that. Search a week signal and correct the tuning of all circuits in the reaciving unit for maximum performace. **This unit needs a shielded box.**



If you keep all values of the coils and capacitors as suggested then you can preset the following settings of the variable-capacitors before you start tuning.

C101: 95% (about 40pF)

C102: 50% (about 22pF)

C103: 33% (about 16pF)

C104: 55% (about 50pF)

C105: 50% (about 35pF)

C106: 50% (about 35pF)

C107: depends on quartz used in the LO

Part list for the converter and lokal oscillator.

The values of the tuning capacitors in the part list are maximum values.

C101, C102, C103, C107 plastic capacitor 7mm, 45pF (violett)

C104 plastic capacitor 10mm, 90pF (red)

C105, C106 plastic capacitor 10mm, 70pF (yellow)

EQ101 quartz 22,0000 MHz

L101 12 turns of silver coated 1mm copper wire wound on a 6mm drill shaft, tapered at the 2 and 5 turn from the cold side.

L102, L103 15 turns of 0.8mm enameld copper wire, wound on a 6mm drill Shaft

L104 18 turns of 0.8mm enameld copper wire, wound on a 6mm drill Shaft

L105 16 turns of 0.8mm enameld copper wire, wound on a 8mm drill Shaft

L106 16 turns of 0.8mm enameld copper wire, wound on a 8mm drill Shaft
Taperd at 3.5 turns from cold side

LDR101 2 x 4 turns of 0.25mm enameld copper wire, wound through a broad-Band ferritbead (binocular)

VDZ zener diode 9.1 V

VT101 FET BF 256

VT102 MOSFET BF 961

VT103 BF 199

Part list for the transmitting converter.

C201, C202, C203 plastic capacitor 7mm, 45pF (violett)

C204 plastic capacitor 10mm, 60pF (yellow)

C205 plastic capacitor 7mm, 60pF (black)

C206 plastic capacitor 10mm, 110pF (violett)

L201 12 turns of 1mm silvercoated copper wire, wound on a 10mm drill-shaft tapered at the center.

L202 9 turns of 1mm silvercoated copper wire, wound on a 10mm drill-shaft

L203 10 turns of 1mm enameld copper wire, wound on a 8mm drill-shaft

L204 8 turns of 1mm enameld copper wire, wound on a 8mm drill-shaft

LDR 201 5 turns of 0.2mm enameld copper wire through a ferrit bead
and 202

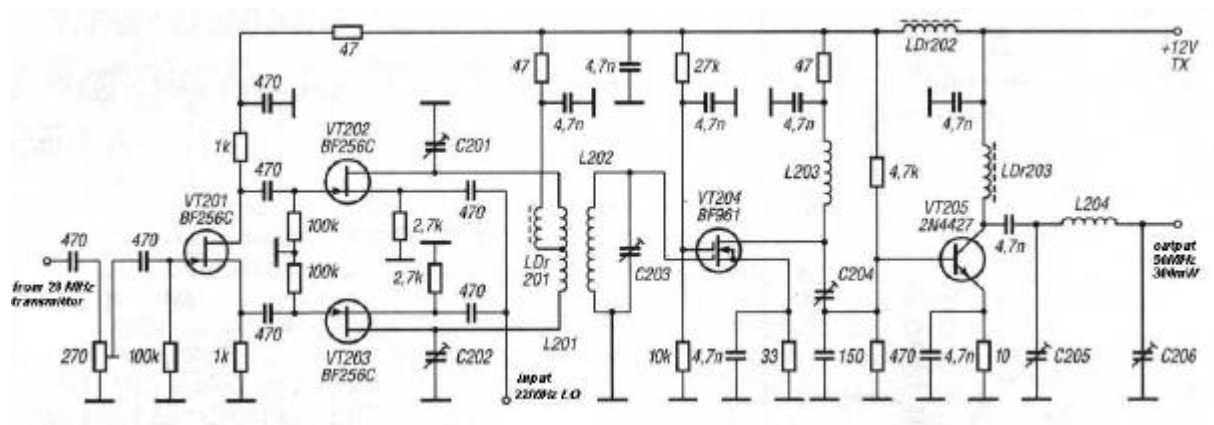
LDR203 2 x 4 turns of 0.25mm enameld copper wire, wound through a broad-Band ferritbead with 2 holes (VHF-choke)

VT201, VT202, VT203 FET BF 256C or BF245C please choose only one type!
VT204 MOSFET BF 961
VT205 2N4427

Look at the scematic for values of the other parts (resistors and capacitors)

All resistors are 0.25 W carbon type. (don't use wire wound ones!!)

All capacitors are cheramik disk type.



Transmitting converter

Even here, (picture 6), I choosed a easy and reliable scematic consept, without the use of a diode ring mixer or any thing similare to it because of costs and coplexity, all the parts together on this PCB are cheaper than one intigrated ring-mixer.

If you don't overload the input of the mixing-unit, you will get a clean signal without harmonics and spourious products. If you wanted to use a IE-500 (ring-mixer) you would need a much higher LO-signal level and one more amplification stage to get it properly working.

VT201 works as a phase inverter unit in order to provide VT202 and VT203 which are setup as a push-pull stage mixer. The 28MHz (26MHz) signal is present to the gate of both transistors, and the LO-signal is added to the sorces. The advantage is, feeding the signals directly without the use of front-end resonant circuits. You only get a clean output signal of 50Mhz when the LO and exciter signals are sucesfully supressed. This is the case when you use two identicall transistors (VT202 and VT203).

The signal then passes trough a bandpass-filter for 50MHZ, to a MOSFET BF 961 (VT204) And is then added after a resonant circuit and capacitiv divider to the driver stage. The driver transistor (VT205) 2N4427 provides enough power, about 150mW, over a tunable Pi-filter to the next stage. (to be continued)

If you download the zip-file all PCB-Layouts and this translation are included.